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DETECTION OF STATIC ECCENTRICITY FAULT IN SATURATED INDUCTION MOTORS BY AIR-GAP MAGNETIC FLUX SIGNATURE ANALYSIS USING FINITE ELEMENT METHOD

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ABSTRACT

Unfortunately, motor current signature analysis (MCSA) cannot detect the small degrees of the purely static eccentricity (SE) defects, while the air-gap magnetic flux signature analysis (FSA) is applied successfully. The simulation results are obtained by using time stepping finite elements (TSFE) method. In order to show the impact of magnetic saturation upon the diagnosis of SE fault, the analysis is carried out for saturated induction motors. The index signatures of static eccentricity fault around fundamental and PSHs are detected successfully for saturated motor.

Key Words: Induction motor, static eccentricity (SE), time stepping finite elements (TSFE) method, air-gap magnetic flux signature analysis (FSA), magnetic saturation.

1. INTRODUCTION

The unbalanced magnetic pull (UMP) is a very important issue in induction motors, the UMP is caused by unequal air gap which called eccentricity fault [1-4]. In general, there are two types of air-gap eccentricity, the static air-gap eccentricity (SE) and the dynamic air gap eccentricity (DE), mixture of both forms called mixed eccentricity [5].

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In the ideal case the stator and rotor have the same geometric axis (fig. 1a), in the case of static eccentricity the rotor rotates around its own geometric axis (fig. 1b), which is not the geometric axis of the stator, in case of dynamic eccentricity (fig. 1c); the rotor is not concentric and rotates around the geometric axis of the stator [3]. However even in new machines eccentricity usually occurs due to the build-up of tolerances during manufacturing stage [6].

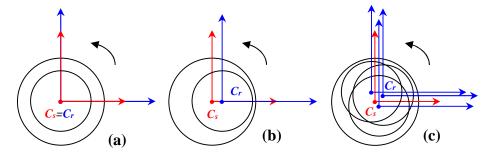


Fig. 1. Types of eccentricity faults, (a) concentric rotor, (b) static eccentricity, (c) dynamic eccentricity

In induction motors the majority of faults are connected to the mechanical and rotor faults, while 80% of mechanical faults result in eccentricity [7]-[8]. The detection of incipient eccentricity faults in induction motors is one of the most important objectives in the maintenance of industrial systems.

Traditionally the Motor Current Signature Analysis (MCSA) is used to diagnose all of the trouble types in induction motors. The initial work in the detection of eccentricity taking account rotor slot related formulas using MCSA and vibration was done by Cameron [9], since there many research works head to detect the eccentricity fault in induction motor using MCSA [10]-[11]-[12]. By another eye, several research works affirmed that MCSA has a limited usability to detect faults in induction motor especially in some cases. Usually MCSA is used to detect a high levels of both eccentricity types, it mean when the static and dynamic eccentricity exist together, but it is very difficult to detect such faults if they appear individually [13]-[5], for those authors the fault signatures are presented in stator current spectrum only if the two kinds of eccentricity exist together and it is not easy to distinguish them. In addition, it will be difficult to diagnose the low level static eccentricity with traditional MCSA technique, for them other signals such as noise, bearing vibration and temperature may have to be used in these cases for detecting fault signature [3]. For Concari et al, diagnostic procedures based only on sideband current components may fail in some

cases for them other induction motor quantities help to detect signatures produced by the different faults such as stray flux and vibrations [14]. Many research works head to use others induction motor quantities to detect SE fault signatures such air gap magnetic flux density [13]-[15], terminal voltages of induction machines at switch-off [5].

Magnetic saturation phenomena is an invariant parameter induction motors, several work research investigate that the magnetic saturation is necessary to study and analyze induction motor suffered from eccentricity fault [16]-[7], while most of induction motor models neglected it. In spite of its inconveniences of a long time of calculation and a need of powerful computers, the power and versatility of the FEM are by now well known and have led to rapid growth in its use. FEM is able to closer the reality and extend accurate results, also unlike other methods FEM offers a high space harmonics of induction motor which open a restful way to diagnose and study the induction motor using signature analysis.

The air gap magnetic field contains full information of the stator condition and mechanical parts of the induction motor, the Fast Fourier Transform (FFT) application to air gap magnetic flux density is used to detect different fault in induction motors, the approach is known as flux signature analysis (FSA). The usability of analysis of external magnetic field to detect faults in induction motor have proved by several papers, the majority of those papers are interested on diagnosis of broken rotor bar [17]-[18]-[14]-[15], stator winding short-circuits [19, 20], net voltage asymmetry [21] wound rotor phase disconnection [22], dynamic eccentricity [13].

This paper heads to diagnose low levels of purely SE in induction motor using TSFE method, short description of this technique modeling will be presented in first part of paper. The application of FSA to detect purely SE in induction motor for constant and non linear permeability is included in the second part of paper.

2. MODELLING OF INDUCTION MOTOR USING TSFE METHOD

In 1988 the Time Stepping Finite Element (TSFE) technique proposed by researchers of GEC research center [23, 24]. Several authors head to the application of this technique for studying and analyzing eccentricity fault in induction motors [10], for them the assumption that currents and fluxes are sinusoidal varying restrict the solution if information such as stator current harmonics and torque pulsation are required, the alternative would be the full time-stepping. Several papers have been used TSFE method to diagnose different faults in induction motors, most of these papers are focused on the diagnosis of broken rotor bar [25]-[26-28], in case of eccentricity analysis [10, 29, 30], in case of stator winding turn fault [31].

TSFE analysis of induction motor based circuit-coupled method

Generally the induction motor is excited by an external voltage source, from many models based on finite elements method, only the Circuit-Coupled Finite Element Method (Circuit-Coupled FEM) can close this reality, here the external electric circuit which contains the three voltage sources is coupled to the finite element domain which contains the magnetic circuit of the electric machine (fig. 2) [32]-[28]. In table 1, specifications of the studied induction motor.

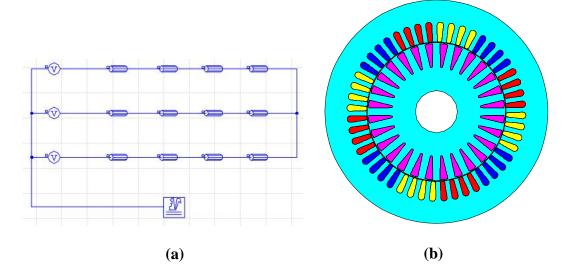


Fig. 2. Circuit coupled FEM mode: a electric circuit; b magnetic circuit

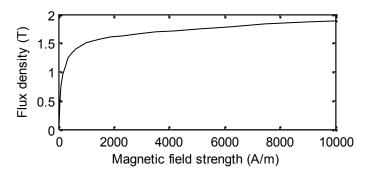


Fig. 3. Magnetization characteristic of core

An initial iron relative permeability of 1000 is assigned to the laminated iron core. For a comparison the calculation are also done for a saturated ferromagnetic material for which the laminated iron region is modeled by a *B*-*H* curve (fig. 3).

Number of poles	4
Number of phases	3
Rated power (<i>kW</i>)	1.1
Rated voltage (V)	230
Frequency (<i>Hz</i>)	50
Rated speed (rpm)	1425
Number of stator slots	36
Number of rotor slots	28

Table 1. Specifications of studied induction motor

For Preston [23] the transient magnetic field in terms of magnetic vector potential A, \in the reluctivity, conductivity \dagger , and current density J can be expressed as:

$$\frac{\partial}{\partial x} \left(\in \frac{\partial A}{\partial x} \right) + \frac{\partial}{\partial y} \left(\in \frac{\partial A}{\partial y} \right) = -J + \dagger \frac{\partial A}{\partial t}$$
(1)

The last term represents current induced in conducting material when flux changes with time. For Salon [33] and Lombard [34] the full transient performance can be obtained by accounting for the mechanical equation of the motor and the transient motional induced eddy current in the rotor bars can't be neglected as in equation (1), the complete system is based on the differential equation in compact form:

$$\in \times \left(\in \nabla \times A \right) = \dagger \left[-\frac{\partial A}{\partial t} - \nabla E + V \times \left(\in \times A \right) \right]$$
 (2)

Where V is the speed of the moving parts, E is the electrical scalar potential applied by the external circuit. So the problem is considered as transient alike. When the motor is loaded the voltage time step must be accompanied by rotor motion corresponding to the slip. The method is direct one if the iron nonlinearity is incorporated in the transient analysis and the external voltage sources are implied within the system of finite element set of simultaneous equation.

Static eccentricity fault diagnosis using air gap magnetic flux density signature analysis In this paper the induction motor under low levels of purely SE (5 and 10 %) is modeled with high precision using TSFE method. In the finite element model the geometry of induction motor has two different centers; stator and rotor centers. Ideally the two centers are placed at the same position, so the air gap is symmetrical, by shifting the rotor center away from the stator center the space-fixed minimum air gap will be formed, this technique is used to model the SE in induction motor.

Frequencies given by the following relationship:

$$f_h = \left[\left(kR \pm n_d \right) \frac{\left(1 - s \right)}{p} \pm \} \right] f_s \tag{3}$$

are used to detect eccentricity-related fault signatures in the stator current, the vibrationrelated spectrum and air gap magnetic flux density. Here $n_d = 0$ in case of static eccentricity, $n_d = 1, 2, 3, ...$ in case of dynamic eccentricity (n_d is known as eccentricity order), f_s is the fundamental supply frequency, R is the number of rotor slots, s is the slip, p is the number of pole pairs, k is any integer and $\}$ is the order of the stator time harmonics that are present in the power supply driving the motor ($\} = \pm 1, \pm 3, \pm 5,$)[5].

It was shown that some of research works leaned on the amplitude variations and the appearing of new components around which called principal slot harmonics (*PSHs*) to diagnose the eccentricity fault in induction motor [35]. *PSHs* are given by the following relationships [36]:

The lower (first) principal harmonic:

$$f_{PSH1} = f_s \left(1 - \frac{R}{p} (1 - s) \right) \tag{4}$$

The upper (second) principal harmonic:

$$f_{PSH2} = f_s \left(1 + \frac{R}{p} \left(1 - s \right) \right) \tag{5}$$

Obviously, the frequencies given by (4) and (5) depend on machine load, by other words; position of *PSHs* in stator current spectrum depends on rotor speed. However, the existence of *PSHs* in current line spectrum depends on the number of pole pairs and rotor slots which lead that the detection of eccentricity fault in stator current spectrum around the *PSHs* is effective only for some combination of number of pole pairs and rotor slots. As will be shown, for modeled induction motor, both *PSHs* exist in stator current and air gap magnetic flux density spectrums. More details of combination of number of pole pairs-rotor slots are presented in [5]-[38]-[37]-[36]. However for all machines static and dynamic eccentricity-related components result around the fundamental harmonic is given by [5]:

$$f_{ecc} = \left| f_s \pm k f_r \right| \tag{6}$$

$$f_r = \frac{\left(1-s\right)}{p} f_s \tag{7}$$

Where k = 1, 2, 3, ...

Figure 4a, b, c presents the magnetic flux distribution, the flux line maps are taken from healthy and faulty saturated induction motor under 10% of SE at starting operation and steady state operation it can be observed that when static eccentricity occurs, the magnetic flux distribution is deformed.

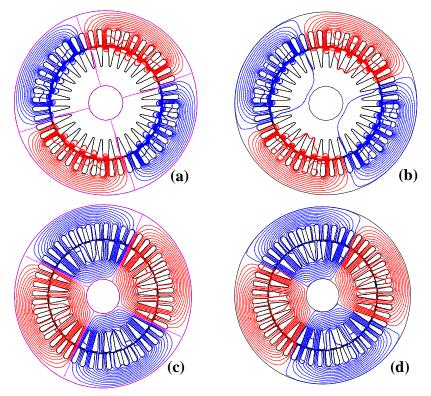


Fig. 4. Flux lines distribution: a starting operation of healthy motor; b starting operation of faulty motor with 10% SE; c steady state operation of healthy motor; d steady state operation of faulty motor with 10% SE

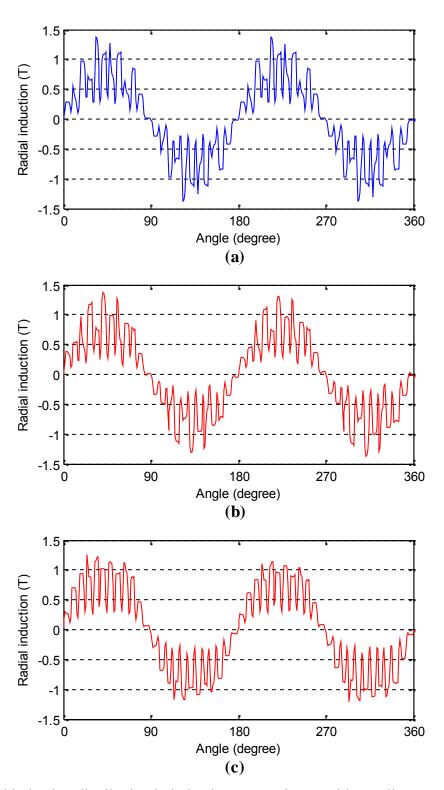


Fig. 5. Radial induction distribution in induction motor air gap with non-linear magnetization characteristic: **a** healthy at full-load; **b** with 10% SE at full-load; **c** with 10% SE at no-load

• Spectrum analysis of air gap magnetic flux density of healthy and faulty motor

Figure 6, presents the air gap flux density spectrums for full-load healthy induction motor under non-linear permeability, it can clearly shown that the *PSH1* and *PSH2* occur

respectively at $((1-0.048/2)28\pm1)50Hz = 616.4Hz$ and 716.4 Hz, also the air gap magnetic flux density spectrum under saturated motor is very rich by the expectation harmonic

components, such saturation harmonics and saturation related harmonics.

Figure 7 shows the air gap magnetic flux density spectrum for full-load faulty induction motor under non-linear permeability, the index signatures of SE fault are successfully appeared around fundamental harmonic and *PSHs*.

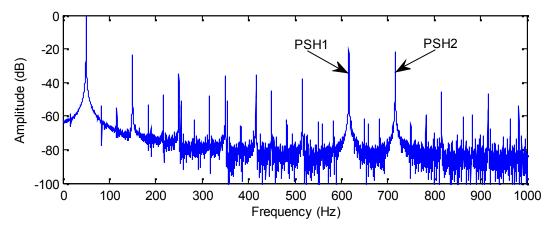


Fig. 6. Air gap magnetic flux density spectrum for full-load healthy induction motor under

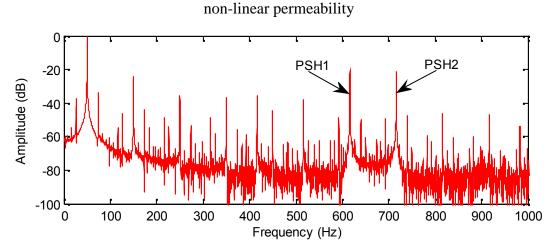


Fig. 7. Air gap magnetic flux density spectrum for full-load faulty induction motor with 10% SE under non-linear permeability

• Static eccentricity index signatures around fundamental harmonic

Figure 8a, b, c shows the spectrums of the air gap magnetic flux density around fundamental harmonic of healthy and faulty saturated induction motor. It can clearly observed that the components at frequency $f_s \pm kf_r$ are successfully appeared. For k = 1, the specific fault harmonics occur at $f_s \pm f_r = 26.2Hz$ and 73.8Hz. As shown in fig. 7b with 5% of SE and fig.

7c with 10% of SE, the amplitude of low side-band component at frequency $f_s - f_r$ increases from -42.54*dB* to -37.02*dB*, also the amplitude of upper-side-band component at frequency $f_s + f_r$ increases from -39.32*dB* to -33.35*dB*. Also as shown in fig. 7c the severity of SE fault with 10% leads to generate the side-band component frequency at $f_s + 2f_r = 97.6Hz$.

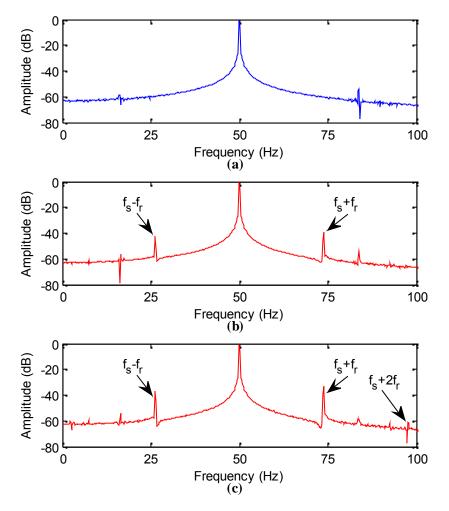


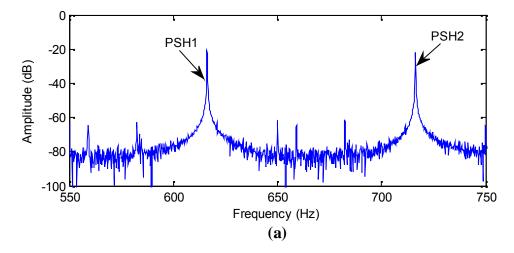
Fig. 8. Zoom around fundamental harmonic of the air gap magnetic flux density spectrum at full-load saturated motor: **a** healthy motor; **b** with 5% of SE; **c** with 10% of SE

Saturated motor				
	$f_s - f_r$	$f_s + f_r$		
5 % SE at Full-load	-42.54	-39.32		
5 % SE at half-load	-43.05	-38.86		
5 % SE at no-load	-42.34	-39.39		
10 % SE at Full-load	-37.02	-33.35		
10 % SE at half-load	-36.55	-32.56		
10 % SE at no-load	-36.40	-33.37		

 Table 2. Amplitude of side-band components around fundamental harmonic for 5% and 10%
 of SE and different loads in *dB*

• Static eccentricity index signatures around PSHs

In addition of SE fault index signatures around fundamental harmonic, fig. 9a, b, c presents the frequency spectrums of the air gap magnetic flux density around *PSHs* for saturated healthy and faulty induction motor, which indicate the sensibility of the air gap magnetic flux density and its high ability to detect a low levels of SE fault.



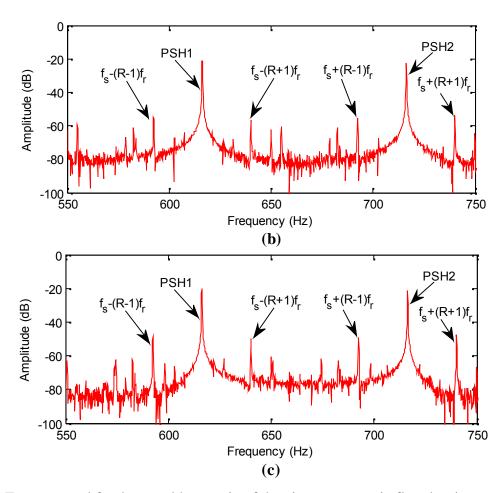


Fig. 9. Zoom around fundamental harmonic of the air gap magnetic flux density spectrum at full-load, saturated motor: **a** healthy motor; **b** with 5% of SE; **c** with 10% of SE.

For full-load motor the frequency components around *PSH1* occur at $f_s - (R \pm 1) f_r = 592.6Hz$ and 640.2Hz, while the frequency components around *PSH2* occur at $f_s + (R \pm 1) f_r = 692.6Hz$ and 740.2Hz. Table 3 summarize the amplitudes and frequencies of components around *PSHs*. The clearly observation that the amplitudes of sideband components increase when the degree of SE fault doubled.

saturated motor					
	Around PSH1		Around PSH2		
	$f_s - (R-1)f_r$	$f_s - (R+1)f_r$	$f_s + (R-1)f_r$	$f_s + (R+1)f_r$	
5 % SE	-55.02	-56.43	-55.44	-53.84	
10 % SE	-48.54	-49.60	-48.90	-47.20	

Table 3. Amplitude of side-band components around PSHs for 5% and 10% of SE at full-
loaded motor in dB

3. CONCLUSION

In this paper, a highly precise modeling TSFE method is used to diagnose low levels of purely SE fault in three phase saturated induction motor. The failure of MCSA to detect the index frequencies of purely SE is successfully replaced by the efficiency and sensibility of the air gap magnetic flux density signature analysis, where the basic known index of SE around fundamental and *PSHs* were appeared very clearly even under very small degree of SE.

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