Performance Evaluation of Wavelet De-Noising Schemes for Suppression of Power Line Noise in Electrocardiogram Signals



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ABSTRACT: Power line noise introduces distortions to recorded electrocardiogram (ECG) signals. These distortions compromise the integrity and negatively affect the interpretation of the ECG signals. Despite the fact that the amplifiers used in biomedical signal processing have high common mode rejection ratio (CMRR), ECG recordings are still often corrupted with residual Power Line Interference (PLI) noise. Further improvement in the hardware solutions do not have significant achievements in PLI noise suppression but rather introduce other adverse effects. Software approach is necessary to refine ECG data. Evaluation of PLI noise suppression in ECG signal in the wavelet domain is presented. The performance of the Hard Threshold Shrinkage Function (HTSF), the Soft Threshold Shrinkage Function (STSF), the Hyperbola Threshold Shrinkage Function (MGTSF) for the suppression of PLI noise are evaluated and compared with the aid of an algorithm. The optimum tuning constant for the Modified Garrote Threshold Shrinkage Function (MGTSF) is found to be 1.18 for PLI noise. GTSF is found to have best performance closely followed by MGTSF in term of filtering Gain. HTSF recorded the lowest Gain. Filtering against PLI noise in the wavelet domain preserves the key features of the signal such as the QRS complex.

KEYWORDS: Electrocardiogram, power line noise, biomedical amplifier, wavelet transformation, wavelet threshold, shrinkage functions.

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I. INTRODUCTION

Bio-electric signals have been very useful in diagnostics and physiological monitoring applications (McSharry et al., 2003; Opie, 2004). Electrocardiogram (ECG) is among the most utilised bio-signals as it is a representation of the electrical activity of the heart in graphical form (Castillo et al., 2013). This activity depicts the action potential, invoked by muscular contraction and relaxation of the heart which varies with prevailing health condition (Dandapat et al., 2015). This information is useful for the detection of various types of heart disorder (Dandapat et al., 2015).

The ECG signal has important diagnostic features marked with hills and valleys as shown in Figure 1. These features are the P-wave, QRS-complex, T-wave, and U-wave. The QRScomplex indicates ventricular contraction while T-wave indicates ventricular relaxation. The U-wave, though not common, indicates delayed purkinje fibres depolarization (Opie, 2004; Dandapat et al., 2015).

The ECG signal S(n) may be corrupted with noise signals $\eta(n)$ during acquisition as electrodes are conventionally attached to the human skin to pick the signals (Romero et al., 2012; Akwei-Sekyere, 2015; Zubair et al., 2018). This is

illustrated in Eqn. (1) and Figure 2 (Zubair et al., 2018). The noise signals include PLI noise introduced by a 50/60Hz AC mains supply, the baseline wander noise generated by lead placement on bone, the rhythmic movement of the chest while breathing and restless movement, and the Electromyography (EMG) noise on the other hand is introduced by the high frequency muscle activities (Zubair et al., 2018; Aqil et al., 2017; Ravandale and Jian, 2015).

$$X(n) = S(n) + \beta \eta(n) \tag{1}$$

where X(n) is the corrupted signal; β is the noise amplification factor.

The PLI noise or simply Power line noise is a significant artefact in the ECG signal. PLI has a band width of less than 1 Hz; It is a narrow-band noise centred at the mains supply frequency of 50Hz or 60Hz (Castillo et al., 2013). Hence, PLI interferes with the ECG signal which has a frequency range of 0.5 Hz - 150 Hz (Aqil et al., 2017). Power line noise introduces distortions to recorded ECG signals. These distortions compromise the integrity and negatively affect the interpretation of the ECG signals (Akwei-Sekyere, 2015). Suppression of Power line noise is necessary in bioengineering, neural engineering, neurosurgery, cardiology and drug discovery (Akwei-Sekyere, 2015).



Figure 1: Typical ECG signal waveform (McSharry et al., 2003).



Figure 2: Dynamics of ECG Signal Corruption with noise (Zubair et al., 2018).

Many approaches have been developed for PLI suppression in ECG signals. The hardware approaches include the use of biomedical amplifiers with amplified CMRR. Despite high CMRR, ECG data are diffused with residual power-line interference due to differences in the electrode impedances. Hence, a false differential signal is created such that an infinitely high CMRR cannot suppress (Chavdar et al., 2005). PLI interferes with the correct boundaries and features of the ECG leading to inaccurate ECG analysis and diagnosis (Chavdar et al., 2005). Further improvement in the hardware solutions do not have significant achievements in PLI noise suppression but rather introduce other adverse effects (Chavdar et al., 2005; Lin and Hu, 2007); Hence, the need for the introduction of software based signal processing solutions (Chavdar et al., 2005). Notable among these software approaches are the window filtering methods such as the rectangular window filtering method which are computationally easy but suboptimal (El Bouny et al., 2019; Alfaouri and Daqrouq, 2008) and the subtraction procedure.

Empirical Mode Decomposition (EMD), being a data driven technique was implemented by Suchetha et al. (2017) to filter 50Hz noise from ECG signal; however, EMD is not robust to small perturbation (Aqil et al., 2017). Numerous adaptive filtering techniques such as the Least Mean Square (LMS), normalized LMS and Recursive Least Square (RLS) among others have been used by Romero et al. (2012) in suppressing PLI noise from ECG signal. Dwivedi et al. (2020) used ensemble approach in ECG noise reduction. Although, ensemble approach yielded a good result, it was

computationally complex. Statistical approach such as the Independent Component Analysis (ICA) has also been used by (Akwei-Sekyere, 2015). Genetic algorithms and artificial intelligence technique have also been applied to ECG noise removal (Mateo et al., 2008). In recent time, wavelet transform technique has been increasingly explored in PLI noise suppression with good results (Chmelka, 2005). Wavelet technique preserves the ECG morphological features which are essential for accurate diagnosis (Tiwari and Dubey, 2014; Huimin et al., 2012). El-Bouny et al. (2017) used undecimated wavelet transform to remove PLI noise from ECG signal while Castillo et al. (2013) used a one-step wavelet processing technique to reduce artefacts in ECG signal. Stationary wavelet transform coupled with interval threshold technique was proposed by (El-Bouny et al., 2017; El-Bouny et al., 2018) to reduce ECG affiliated noise.

In this work, suppression of PLI noise in wavelet domain was studied with the help of an algorithm developed in MATLAB working environment. The performance of the Hard Threshold Shrinkage Function (HTSF), the Soft Threshold Shrinkage Function (STSF), the Hyperbola Threshold Shrinkage Function (HYTSF), the Garrote Threshold Shrinkage Function (GTSF), and the Modified Garrote Threshold Shrinkage Function (MGTSF) for the suppression of PLI noise are evaluated and compared.

II. THEORETICAL BACKGROUND

A. Materials

Signal transformation is usually done to simplify signal processing. The Fourier transform, being a very popular transform can only present the frequency components of a signal that is stationary (Zubair and Ahmed, 2019; El Bouny et al., 2018; Aqil et al., 2015; Goswami and Chan, 2011). However, physiological signals are non-stationary and can best be processed in wavelet transform (Goldberger et al., 2000). Wavelet transform can present the non-stationary signals simultaneously in time and frequency domains and is widely adopted in signal noise suppression or compression tasks (Zubair, 2019).

The energy of a small wave (wavelet) is concentrated in time (Zubair and Ahmed, 2019). Wavelet families include Coiflets, Spline, Haar, Daubechies (Db) among others. Figure 3 illustrated the wavelet decomposition of a corrupted signal in to two parts with L levels. Part A is the approximate coefficients containing the low-frequency information of the signal while part D is the details components of the coefficients and it contains information about the high-frequency part of the signal. Low pass (lp) and high pass (hp) filters are used to generate A and D respectively. Given that Y is the wavelet transform of X. The approximate component Ya(n) of Y and Detail component Yd(n) of Y are related to A and D as indicated in Figure 3 (Zubair et al., 2018; Van Fleet, 2011; Donoho and Johnstone, 1994).

The Threshold Shrinkage functions considered in this work are the HTSF, STSF, HYTSF, GTSF and MGTSF which are defined by Eqns. (2), (3), (4), (5), and (6) respectively (Zubair et al., 2018; Van Fleet, 2011). σ , t and λ are defined as the noise level, universal threshold and modified universal threshold respectively and are given by Eqns. (7), (8), and (9) respectively (Zubair and Ahmed, 2019; Van Fleet, 2011; Donoho and Johnstone, 1994;). To achieve noise suppression, the universal threshold t proposed by Donoho and Johnstone (1994) also called Visu shrink is adopted. A modified universal threshold λ which is a product of the universal threshold t and a tuning constant (α) as shown in Eqn. (9) is used for the Modified Garrote Threshold Shrinkage Function (MGTSF).



 Y_{ma}

Figure 3: A three level wavelet decomposition diagram.

B. Wavelet Shrinkage Function Schemes

The overall block diagram of the noise suppression process of wavelet transform using threshold shrinkage function is shown in Figure 4. The input to the scheme is a 1 by 2N sequence corrupted ECG signal X. Discrete Wavelet decomposition (DWT) at level L, transforms X to a 1 by 2N sequence, Y (Zubair and Ahmed, 2019; Van Fleet, 2011; Donoho and Johnstone, 1994). This Y is sorted into Approximation coefficients Ya and Detail Coefficients Yd. while Yd is modified and shrank by relevant functions to become Ydm, Ya is preserved. In the end, Ya and Ydm are recombined to yield the wavelet transformed coefficients, Yf. Using the Inverse Discrete Wavelet Transform at level L, Yf is inverse-transformed to Xf (Zubair and Ahmed, 2019; Van Fleet, 2011; Donoho and Johnstone, 1994). The recovered signal after the filtering process, Xf is expected to be similar in structure and content to the original ECG signal S of Figure 2. Features such as the QRS complex and R-R interval are preserved by this scheme.



Figure 4: ECG Wavelet threshold shrinkage function de-noising scheme.

In order to achieve noise suppression, a Threshold Shrinkage Function is used to modify the detail part Yd to give modified detail part Ymd. Any element in Yd which is equal to or less than a threshold is set to zero while an element which is greater than the threshold is modified.

$$_{i}(i,j) = \begin{cases} Y_{d}(i,j) & \text{ for } |Y_{d}(i,j)| > t \\ 0 & \text{ otherwise} \end{cases}$$
(2)

$$Y_{md}(i,j) = \begin{cases} \operatorname{sgn}[Y_d(i,j)](Y_d(i,j)|-t) & \operatorname{for}|Y_d(i,j)| > t \\ 0 & \operatorname{otherwise} \end{cases}$$
(3)

$$Y_{md}(i,j) = \begin{cases} \operatorname{sgn}[Y_d(i,j]\sqrt{\left[Y_d(i,j)\right]^2 - t^2}) & \operatorname{for}|Y_d(i,j)| > t \\ 0 & \operatorname{otherwise} \end{cases}$$
(4)

$$Y_{md}(i,j) = \begin{cases} Y_d(i,j) - \frac{t^2}{Y_d(i,j)} & \text{for} |Y_d(i,j)| > t \\ 0 & \text{otherwise} \end{cases}$$
(5)

otherwise

$$Y_{md}(i,j) = \begin{cases} \frac{\left[Y_d(i,j)\right]^2 - (\lambda - \sigma)^2}{Y_d(i,j)} & \text{for} |Y_d(i,j)| > \lambda \\ 0 & \text{otherwise} \end{cases}$$
(6)

$$\sigma = \frac{median\left[Y_d\right]}{0.6745} \tag{7}$$

$$t = \sigma \sqrt{2Log_e(N)} \tag{8}$$

where N is the number of elements in Yd.

$$\lambda = \alpha t \tag{9}$$

This wavelet de-noising scheme performance is evaluated using the Signal to Noise Ratio (SNR) (Kabir and Shahnaz, 2012; Dandapat et al., 2015). SNRc of Eqn. (10) compares the noisy signal X with the original signal S. PSNRf of Eqn. (11) ▶ Xf compares the de-noised signal Xf with the original signal S. Filtered The Gain of Eqn. (12) is positive if there is some degree of noise suppression.

$$SNR_{c} = 10 \log_{10} \frac{\sum_{n=1}^{n=2N} [S(n)]^{2}}{\sum_{n=1}^{n=2N} [S(n) - X(n)]^{2}}$$
(10)

$$SNR_f = 10\log_{10} \frac{\sum_{n=1}^{n=2N} [S(n)]^2}{\sum_{n=1}^{n=2N} [S(n) - Xf(n)]^2}$$
(11)

$$Gain = SNR_f - SNR_c \tag{12}$$

Figure 4 is coded into an algorithm in MATLAB working environment using relevant MATLAB resources (MathWorks, 2019). η is the PLI noise; The noise amplification factor, $\beta = 30$ is used to simulate noticeable mix of the noise with the test signals. The algorithm is summarised in Table 1 below.

Table 1: PLI noise suppression algorithm.

Step

Initialize ECG signal S by extraction from MIT-BIH Arrhythmia database.

Add 50Hz PLI noise to the ECG signal to obtain the corrupted ECG signal X.

Set the level of wavelet decomposition (L=3) and compute Y [Ya:Yd], the wavelet transform of X.

Obtain the noise level, universal threshold and modified universal threshold

Apply a Wavelet Threshold Shrinkage Function to modify Yd to Ymd.

Obtain Yf [Ya:Ymd] and apply the inverse DWT to reconstruct filtered ECG signal Xf.

Initialize ECG signal S by extraction from MIT-BIH Arrhythmia database.

Compute PSNR_c, PSNR_f and the Gain.

III. RESULTS AND DISCUSSION

A. Test Signals

Six ECG signals data were obtained from the MIT-BIH Arrhythmia database. The database has recordings of 48 real ECG data obtained from two leads, sampled at 360 Hz for 30 min duration at 11 bit resolution. ECG signals with data record no: 100.dat, 101.dat, 102.dat, 103.dat, 104.dat, and 105.dat were selected. The first 1000 samples of the data were considered for the analysis (Moody and Mark, 2001; Physiobank, 2018). A is the actual amplitude of the ECG signal and An is the normalised amplitude of the signal. A and An are related as described by Eqns. (13) and (14).

$$A = \frac{(A_n - 200)}{2^{10}} \tag{13}$$

$$A_n = (2^{10})A + 200 \tag{14}$$

B. Tuning constant for the Modified Garrote Threshold Shrinkage Function

To obtain the optimal tuning constant, the algorithm was tested with the six test ECG signals using the Modified Garrote Threshold Shrinkage Function (MGTSF) for different values of tuning constant α . The outcomes are summarised in Figure 5. The average Gain values obtained for the six test ECG signals for each value of the tuning constant are plotted in Figure 6. In Figure 6, the optimum tuning constant with highest average Gain is 1.18 dB.

C. Wavelet Threshold Shrinkage Functions

The algorithm was tested with the six test ECG signals for the other four wavelet Shrinkage Functions namely HTSF, STSF, HYTSF and the GTSF. The results are combined with those obtained for the MGTSF at optimum tuning constant (α =1.18) as presented in Figure 7.



Figure 5: Plots of the gain against the tuning constant for the six test ECG signals with MGTSF.







Figure 7: Gain obtained for five wavelet threshold shrinkage functions for the six ECG test signals.

The values of the average Gain obtained by each of the five Threshold Shrinkage Functions are plotted in the bar chart of Figure 8. It shows that GTSF has the best performance closely followed by MGTSF. HTSF recorded the lowest Gain and Average Gain.

The results in term of the waveforms, SNR_c , SNR_f , and the Gain for the five wavelet Threshold Shrinkage Functions for Test ECG Signal 2 are presented in Figure 9 which illustrates the same trend as in Figure 8. For the test ECG signal 1, the MGTSF has a slightly higher Gain than the GTSF as shown in Figure 10. The GTSF results in term of the waveforms, SNR_c , SNR_f , and the Gain for the test ECG Signals 3 to 6 are presented in Figs. 11 and 12. Observation of the filtered ECG waveforms shows the preservation of the important features of the ECG signal such as the R-R wave peaks and QRS complex.



Figure 8: Average gain obtained for five wavelet threshold shrinkage functions.



Figure 9: GTSF, MGTSF, HYTSF, STSF and HTSF results for the test ECG signal 2.



Figure 10: GTSF and MGTSF results for the test ECG signal 1.



Figure 11: GTSF results for the test ECG signals 3 and 4.



Figure 12: GTSF results for the test ECG signals 5 and 6.

The low Gain recorded for STSF in Figure 7 could be due to its constant deviation and reduction of the high coefficients which caused significant bias. Hence, it is less sensitive to small change in the data. In Figure 9, the HTSF on the other hand tends to distort the ECG signal morphology as it either shrinks or retain coefficients depending on whether it is lower or higher than the universal threshold respectively. HYTSF had moderate performance in terms of signal Gain, fairly better than STSF and HTSF as seen in Figure 7.

IV. CONCLUSION

A wavelet based PLI noise suppression algorithm has been implemented. The performance of the HTSF, STSF, HYTSF, GTSF and MGTSF for the suppression of PLI noise have been evaluated and compared. Filtering of ECG signal in the wavelet domain has been shown to be useful in the removal of PLI noise with better fidelity and higher noise suppression by testing an array of threshold shrinkage functions. Hence, diagnostic accuracy would be enhanced. In order to take this work further, other wavelet families could be employed while performance on wavelet packet tree and stationary wavelet transform (SWT) could be tested.

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